

Problem 4.15:

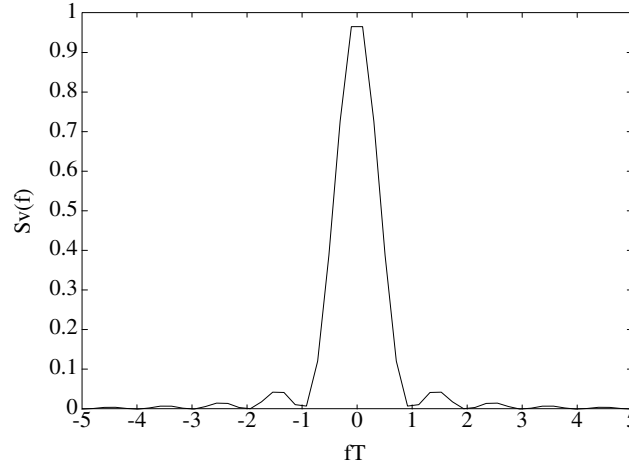
We have that $\Phi_{uu}(f) = \frac{1}{T} |G(f)|^2 \Phi_{ii}(f)$ But $E(I_n) = 0$, $E(|I_n|^2) = 1$, hence : $\phi_{ii}(m) = \begin{cases} 1, & m = 0 \\ 0, & m \neq 0 \end{cases}$. Therefore : $\Phi_{ii}(f) = 1 \Rightarrow \Phi_{uu}(f) = \frac{1}{T} |G(f)|^2$.

(a) For the rectangular pulse :

$$G(f) = AT \frac{\sin \pi f T}{\pi f T} e^{-j2\pi f T/2} \Rightarrow |G(f)|^2 = A^2 T^2 \frac{\sin^2 \pi f T}{(\pi f T)^2}$$

where the factor $e^{-j2\pi f T/2}$ is due to the $T/2$ shift of the rectangular pulse from the center $t = 0$. Hence :

$$\Phi_{uu}(f) = A^2 T \frac{\sin^2 \pi f T}{(\pi f T)^2}$$

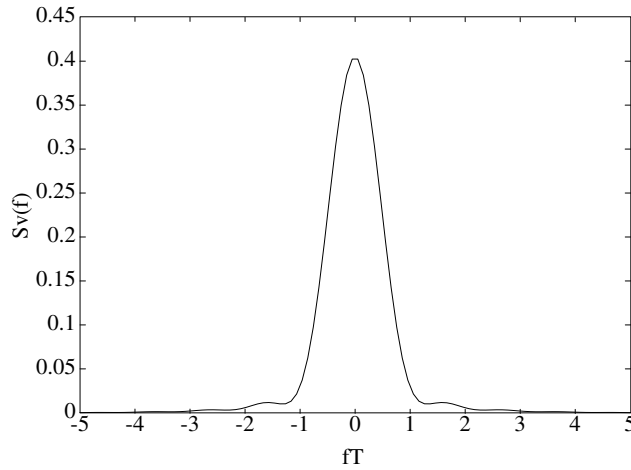


(b) For the sinusoidal pulse : $G(f) = \int_0^T \sin \frac{\pi t}{T} \exp(-j2\pi f t) dt$. By using the trigonometric identity $\sin x = \frac{\exp(jx) - \exp(-jx)}{2j}$ it is easily shown that :

$$G(f) = \frac{2AT}{\pi} \frac{\cos \pi T f}{1 - 4T^2 f^2} e^{-j2\pi f T/2} \Rightarrow |G(f)|^2 = \left(\frac{2AT}{\pi} \right)^2 \frac{\cos^2 \pi T f}{(1 - 4T^2 f^2)^2}$$

Hence :

$$\Phi_{uu}(f) = \left(\frac{2A}{\pi} \right)^2 T \frac{\cos^2 \pi T f}{(1 - 4T^2 f^2)^2}$$



(c) The 3-db frequency for (a) is :

$$\frac{\sin^2 \pi f_{3db} T}{(\pi f_{3db} T)^2} = \frac{1}{2} \Rightarrow f_{3db} = \frac{0.44}{T}$$

(where this solution is obtained graphically), while the 3-db frequency for the sinusoidal pulse on (b) is :

$$\frac{\cos^2 \pi T f}{(1 - 4T^2 f^2)^2} = \frac{1}{2} \Rightarrow f_{3db} = \frac{0.59}{T}$$

The rectangular pulse spectrum has the first spectral null at $f = 1/T$, whereas the spectrum of the sinusoidal pulse has the first null at $f = 3/2T = 1.5/T$. Clearly the spectrum for the rectangular pulse has a narrower main lobe. However, it has higher sidelobes.

Problem 4.20 :

The autocorrelation function for $u_{\Delta}(t)$ is :

$$\begin{aligned}
 \phi_{u_{\Delta}u_{\Delta}}(t) &= \frac{1}{2}E[u_{\Delta}(t+\tau)u_{\Delta}^*(t)] \\
 &= \frac{1}{2}\sum_{n=-\infty}^{\infty}\sum_{m=-\infty}^{\infty}E(I_mI_n^*)E[u(t+\tau-mT-\Delta)u^*(t-nT-\Delta)] \\
 &= \frac{1}{2}\sum_{n=-\infty}^{\infty}\sum_{m=-\infty}^{\infty}\phi_{ii}(m-n)E[u(t+\tau-mT-\Delta)u^*(t-nT-\Delta)] \\
 &= \frac{1}{2}\sum_{m=-\infty}^{\infty}\phi_{ii}(m)\sum_{n=-\infty}^{\infty}E[u(t+\tau-mT-nT-\Delta)u^*(t-nT-\Delta)] \\
 &= \frac{1}{2}\sum_{m=-\infty}^{\infty}\phi_{ii}(m)\sum_{n=-\infty}^{\infty}\int_0^T\frac{1}{T}u(t+\tau-mT-nT-\Delta)u^*(t-nT-\Delta)d\Delta
 \end{aligned}$$

Let $a = \Delta + nT$, $da = d\Delta$, and $a \in (-\infty, \infty)$. Then :

$$\begin{aligned}
 \phi_{u_{\Delta}u_{\Delta}}(t) &= \frac{1}{2}\sum_{m=-\infty}^{\infty}\phi_{ii}(m)\sum_{n=-\infty}^{\infty}\int_{nT}^{(n+1)T}\frac{1}{T}u(t+\tau-mT-a)u^*(t-a)da \\
 &= \frac{1}{2}\sum_{m=-\infty}^{\infty}\phi_{ii}(m)\frac{1}{T}\int_{-\infty}^{\infty}u(t+\tau-mT-a)u^*(t-a)da \\
 &= \frac{1}{T}\sum_{m=-\infty}^{\infty}\phi_{ii}(m)\phi_{uu}(\tau-mT)
 \end{aligned}$$

Thus we have obtained the same autocorrelation function as given by (4.4.11). Consequently the power spectral density of $u_{\Delta}(t)$ is the same as the one given by (4.4.12) :

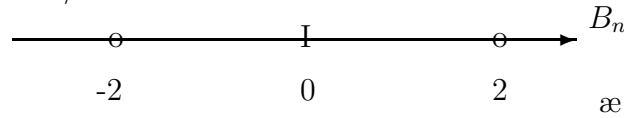
$$\Phi_{u_{\Delta}u_{\Delta}}(f) = \frac{1}{T}|G(f)|^2\Phi_{ii}(f)$$

Problem 4.21 :

(a) $B_n = I_n + I_{n-1}$. Hence :

I_n	I_{n-1}	B_n
1	1	2
1	-1	0
-1	1	0
-1	-1	-2

The signal space representation is given in the following figure, with $P(B_n = 2) = P(B_n = -2) = 1/4$, $P(B_n = 0) = 1/2$.



(b)

$$\begin{aligned}\phi_{BB}(m) &= E[B_{n+m}B_n] = E[(I_{n+m} + I_{n+m-1})(I_n + I_{n-1})] \\ &= \phi_{ii}(m) + \phi_{ii}(m-1) + \phi_{ii}(m+1)\end{aligned}$$

Since the sequence $\{I_n\}$ consists of independent symbols :

$$\phi_{ii}(m) = \begin{cases} E[I_{n+m}]E[I_n] = 0 \cdot 0 = 0, & m \neq 0 \\ E[I_n^2] = 1, & m = 0 \end{cases}$$

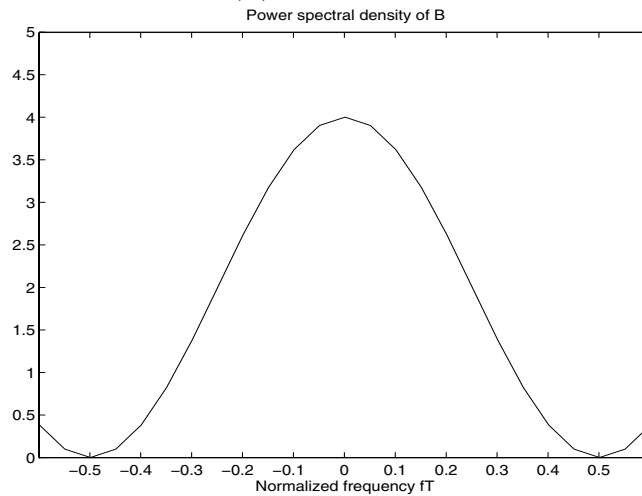
Hence :

$$\phi_{BB}(m) = \begin{cases} 2, & m = 0 \\ 1, & m = \pm 1 \\ 0, & \text{o.w} \end{cases}$$

and

$$\begin{aligned}\Phi_{BB}(f) &= \sum_{m=-\infty}^{\infty} \phi_{BB}(m) \exp(-j2\pi f m T) = 2 + \exp(j2\pi f T) + \exp(-j2\pi f T) \\ &= 2[1 + \cos 2\pi f T] = 4 \cos^2 \pi f T\end{aligned}$$

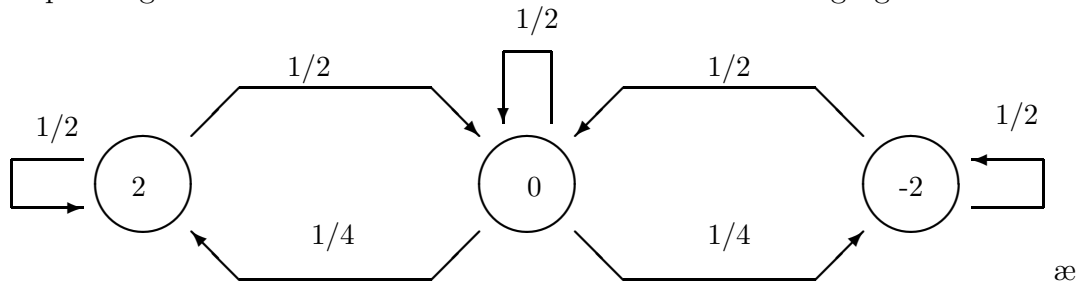
A plot of the power spectral density $\Phi_B(f)$ is given in the following figure :



(c) The transition matrix is :

I_{n-1}	I_n	B_n	I_{n+1}	B_{n+1}
-1	-1	-2	-1	-2
-1	-1	-2	1	0
-1	1	0	-1	0
-1	1	0	1	2
1	-1	0	-1	-2
1	-1	0	1	0
1	1	2	-1	0
1	1	2	1	2

The corresponding Markov chain model is illustrated in the following figure :



Problem 4.22 :

(a) $I_n = a_n - a_{n-2}$, with the sequence $\{a_n\}$ being uncorrelated random variables (i.e $E(a_{n+m}a_n) = \delta(m)$). Hence :

$$\begin{aligned}
 \phi_{ii}(m) &= E[I_{n+m}I_n] = E[(a_{n+m} - a_{n+m-2})(a_n - a_{n-2})] \\
 &= 2\delta(m) - \delta(m-2) - \delta(m+2) \\
 &= \begin{cases} 2, & m = 0 \\ -1, & m = \pm 2 \\ 0, & \text{o.w.} \end{cases}
 \end{aligned}$$

(b) $\Phi_{uu}(f) = \frac{1}{T} |G(f)|^2 \Phi_{ii}(f)$ where :

$$\begin{aligned}
 \Phi_{ii}(f) &= \sum_{m=-\infty}^{\infty} \phi_{ii}(m) \exp(-j2\pi f m T) = 2 - \exp(j4\pi f T) - \exp(-j4\pi f T) \\
 &= 2[1 - \cos 4\pi f T] = 4 \sin^2 2\pi f T
 \end{aligned}$$

and

$$|G(f)|^2 = (AT)^2 \left(\frac{\sin \pi f T}{\pi f T} \right)^2$$

Therefore :

$$\Phi_{uu}(f) = 4A^2T \left(\frac{\sin \pi fT}{\pi fT} \right)^2 \sin^2 2\pi fT$$

(c) If $\{a_n\}$ takes the values (0,1) with equal probability then $E(a_n) = 1/2$ and $E(a_{n+m}a_n) = \begin{cases} 1/4, & m \neq 0 \\ 1/2, & m = 0 \end{cases} = [1 + \delta(m)]/4$. Then :

$$\begin{aligned} \phi_{ii}(m) &= E[I_{n+m}I_n] = 2\phi_{aa}(0) - \phi_{aa}(2) - \phi_{aa}(-2) \\ &= \frac{1}{4} [2\delta(m) - \delta(m-2) - \delta(m+2)] \end{aligned}$$

and

$$\begin{aligned} \Phi_{ii}(f) &= \sum_{m=-\infty}^{\infty} \phi_{ii}(m) \exp(-j2\pi f mT) = \sin^2 2\pi fT \\ \Phi_{uu}(f) &= A^2T \left(\frac{\sin \pi fT}{\pi fT} \right)^2 \sin^2 2\pi fT \end{aligned}$$

Thus, we obtain the same result as in (b) , but the magnitude of the various quantities is reduced by a factor of 4 .

Problem 4.23 :

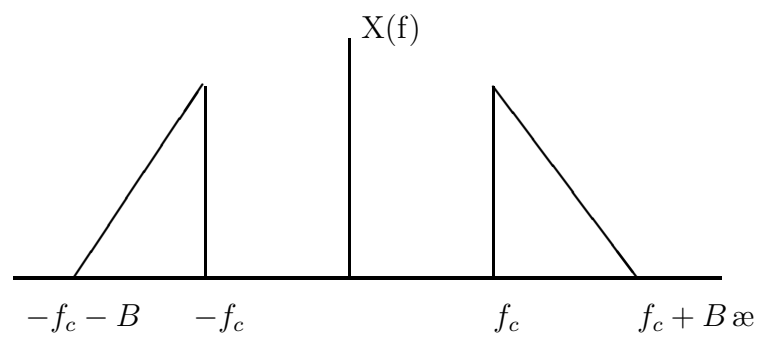
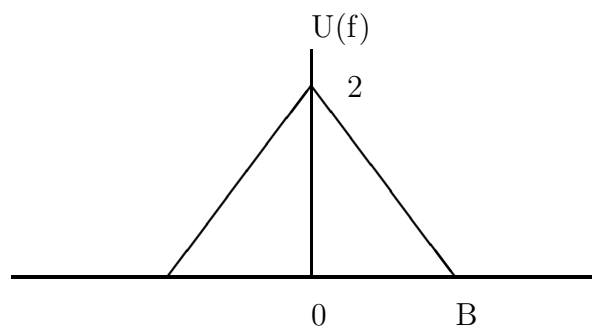
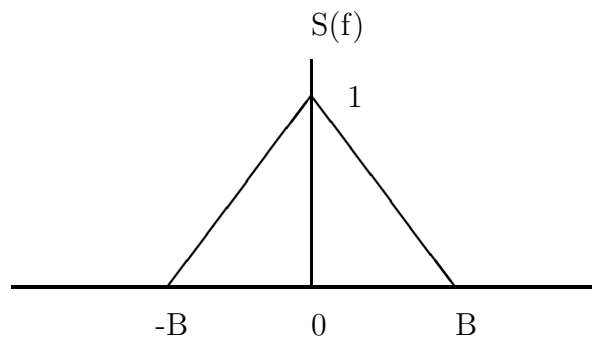
$x(t) = \text{Re} [u(t) \exp(j2\pi f_c t)]$ where $u(t) = s(t) \pm j\hat{s}(t)$. Hence :

$$U(f) = S(f) \pm j\hat{S}(f) \quad \text{where } \hat{S}(f) = \begin{cases} -jS(f), & f > 0 \\ jS(f), & f < 0 \end{cases}$$

So :

$$U(f) = \begin{cases} S(f) \pm S(f), & f > 0 \\ S(f) \mp S(f), & f < 0 \end{cases} = \begin{cases} 2S(f) \text{ or } 0, & f > 0 \\ 0 \text{ or } 2S(f), & f < 0 \end{cases}$$

Since the lowpass equivalent of $x(t)$ is single-sideband, we conclude that $x(t)$ is a single-sideband signal, too. Suppose, for example, that $s(t)$ has the following spectrum. Then, the spectra of the signals $u(t)$ (shown in the figure for the case $u(t) = s(t) + j\hat{s}(t)$) and $x(t)$ are single-sideband



Problem 4.30 :

The 16-QAM signal is represented as $s(t) = I_n \cos 2\pi ft + Q_n \sin 2\pi ft$, where $I_n = \{\pm 1, \pm 3\}$, $Q_n = \{\pm 1, \pm 3\}$. A superposition of two 4-QAM (4-PSK) signals is :

$$s(t) = G [A_n \cos 2\pi ft + B_n \sin 2\pi ft] + C_n \cos 2\pi ft + D_n \sin 2\pi ft$$

where $A_n, B_n, C_n, D_n = \{\pm 1\}$. Clearly : $I_n = GA_n + C_n$, $Q_n = GB_n + D_n$. From these equations it is easy to see that $G = 2$ gives the requires equivalence.